

MICROPOWER 1A LOW DROPOUT PMOS VOLTAGE REGULATOR

FEATURES

- Stable with Ceramic Capacitor
- Small, Space Saving MLPM 6-Pin Package
- Guaranteed < 1V Dropout at Full Load Current
- Fast Transient Response
- Ultra-Low Ground Current
- Output Current Limiting
- Built-In Thermal Shutdown

APPLICATIONS

- High Efficiency Linear Regulator
- Hard Disk Drivers, CD-ROMs, DVDs
- ADSL and Cable Modems

DESCRIPTION

The IRU1502-33 is a PMOS low dropout, linear regulator and it is capable of supplying 1A of continuous current over line and temperature range. The IRU1502-33 is stable with low value ceramic capacitors, ensures low noise operation, improves load transient response and enables a smaller circuit size.

IRU1502-33 features ultra low noise, fast start-up and an excellent time and load response. This device also includes built-in output protection with both current limit and thermal shutdown.

TYPICAL APPLICATION

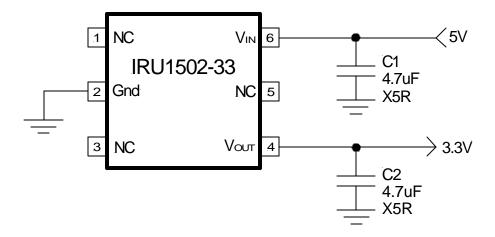


Figure 1 - Typical application of IRU1502-33.

PACKAGE ORDER INFORMATION

T	T _J (°C) DEVICE		PACKAGE	MARKING	
0 T	Го 125	IRU1502-33CH	6-Pin MLPM 3x3 (H)	1502	

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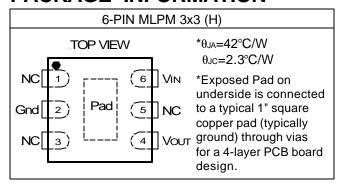
ABSOLUTE MAXIMUM RATINGS

Input Voltage (V_{IN}) 6V

Operating Ambient Temperature Range-40°C To 125°C Storage Temperature Range-65°C To 150°C

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device.

PACKAGE INFORMATION



ELECTRICAL SPECIFICATIONS

Unless otherwise specified, these specifications apply over V_{IN} =4.5V to 5.5V, I_{OUT} =2mA to 1A, C_{IN} =10 μ F, C_{OUT} =10 μ F

PARAMETER	SYM	TEST CONDITION	MIN	TYP	MAX	UNITS
Output Voltage 3.3V	VO(3.3)	4.75V <v<sub>IN<5.25V, 5mA≤lo≤1A:</v<sub>				
		TJ=25°C	3.234	3.3	3.366	V
		0°C≤TJ≤125°C	3.2175	3.3	3.3825	
Line Regulation	RegLINE	4.75V <vin<5.25v, lo="5mA</td"><td></td><td></td><td>15</td><td>mV</td></vin<5.25v,>			15	mV
Load Regulation	RegLOAD	V _{IN} =4.75V, 10mA≤lo≤1A			100	mV
Dropout Voltage	VD	VIN=4V, Io=1A	3		3.3825	٧
		V _{IN} =3.8V, lo=0.8A, Note 2	3		3.3825	
Current Limit	ls	V _{IN} =5.5V	1	1.4		Α
Minimum Output Current	lo(MIN)	Note 3			2	mA
Temperature Stability	Ts	Note 4, 5		0.5		%
RMS Output Noise	Vn	10Hz <bw<10khz, 5<="" note="" td=""><td></td><td>0.003</td><td></td><td>%Vo</td></bw<10khz,>		0.003		%Vo
Ripple Rejection	RA	V _{IN} =5V, f=120Hz, Note 5	45	55		dB
Thermal Shutdown	TJ(SD)	V _{IN} =4.75V, 5mA≤lo≤1A, Note 5	135			°C
Quiescent Current	IGND	Vin≤5.5V, 2mA≤lo≤1A			650	μΑ
Transient Response	ΔVουτ	V _{IN} =5V, Any 200mA step			5	%
Step Load Change (light load	Δ lout	from 100mA to1A, t _r ≥1μs, Note 5				
to full load) Droop Voltage						
Transient Response	ΔVουτ	V _{IN} =5V, 1A to 10mA, t _f ≥1μs, Note 5			3.6	V
Step Load Change (full load to	ΔΙουτ					
light load) Output Voltage						
Transient Response	ΔVουτ	0 to 5V step input, t _r ≥1μs,			3.6	٧
Change of Vout with	ΔVIN	10mA≤lo≤1A				
Application of V _{IN}		Note 5				
Transient Response	ΔVουτ	VIN=5V, Io=ISHORT to Io=10mA			3.6	V
Short Circuit Removal	ΔΙουτ	Note 5				
Response	@IO=Short					

Note 1: Low duty cycle pulse testing with Kelvin connections is required in order to maintain accurate data.

Note 2: In general, Dropout voltage is defined as the minimum differential voltage between V_{IN} and V_{OUT} required to maintain regulation at V_{OUT}. In this specification, it is the measured output voltage at specified condition.

Note 3: Minimum load current is defined as the minimum current required at the output in order for the output voltage to maintain regulation.

Note 4: Temperature stability is the change in output from nominal over the operating temperature range.

Note 5: Guaranteed by design, but not tested in production.

Note 6: All limits are guaranteed. All electrical characteristics have temperature limits that are tested during $T_A=25^{\circ}C$ at probing and tested at final production with $T_A=100^{\circ}C$. All hot and cold limits are guaranteed by correlating the electrical characteristics to process and temperature variations.

PIN DESCRIPTIONS

PIN#	PIN SYMBOL	PIN DESCRIPTION
1,3,5	NC	No connection.
2 4	Gnd	Ground pin.
4	Vout	The output of the regulator. A minimum of 4.7µF output capacitance must be connected
		from this pin to ground to insure stability.
6	Vin	The power input pin of the regulator. A minimum of input capacitance must be connected from this pin to ground to insure that the input voltage does not sag below the minimum dropout voltage during the load transient response. This pin must always be higher than the Vout pin by the amount of dropout voltage (see electrical specification) in order for the device to regulate properly.

BLOCK DIAGRAM

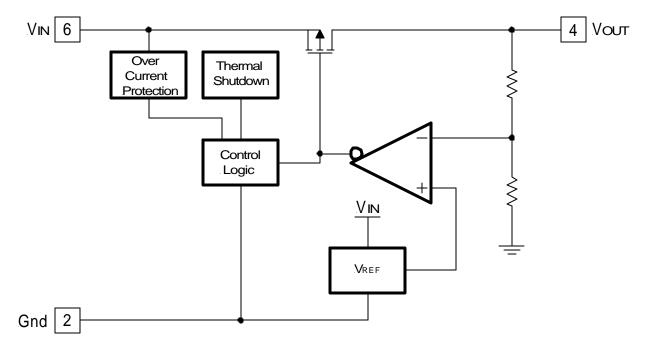


Figure 2 - Simplified block diagram of the IRU1502-33.

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Unless specified, the test data applies: T_A=25°C, C_{IN}=4.7μF, C_{OUT}=4.7μF ceramic and V_{IN}=5V.

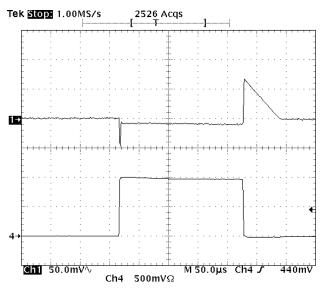


Figure 3 - Step load response from 2mA to 1A, tr≥1μs. Ch1: Output voltage, AC, 50mV/div Ch4: Load Current, 0.5A/div

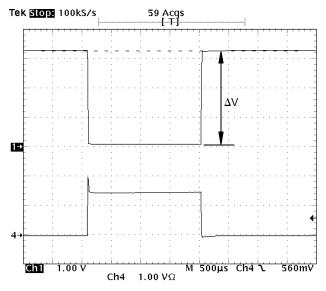


Figure 5 - Output short circuit operation. $\Delta V = 3.26V$ Ch1: Output voltage, 1V/div Ch4: Load Current, 1A/div

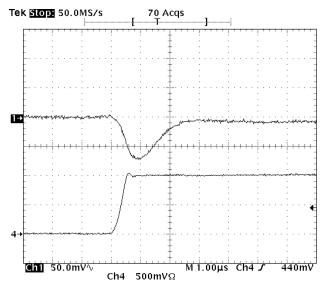


Figure 4 - Step-up transient load response from 2mA to 1A, tr<1µs.
Ch1: Output voltage, AC, 50mV/div
Ch4: Load Current, 0.5A/div

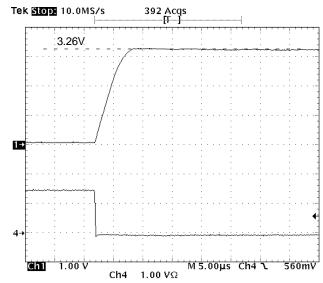


Figure 6 - Short circuit removal, loυτ from short to 10mA. Ch1 Peak: 3.26V Ch1: Output voltage, 1V/div Ch4: Load Current, 1A/div

Unless specified, the test data applies: T_A=25°C, C_{IN}=4.7μF, C_{OUT}=4.7μF ceramic and V_{IN}=5V.

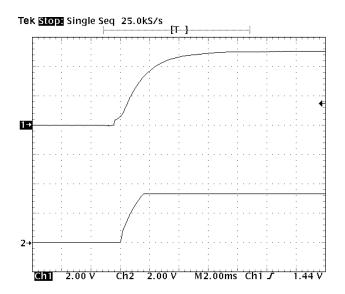


Figure 7 - Start-up at lout=10mA. Ch1: 5V input voltage, 2V/div Ch2: 3.3V output voltage, 2V/div

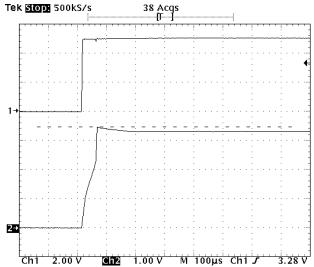


Figure 9 - Input voltage transient response, V_{IN} from 0V to 5V, Couτ=10μF. Ch2 Peak: 3.48V Ch1: 5V input voltage, 2V/div

Ch2: 3.3V output voltage, 1V/div

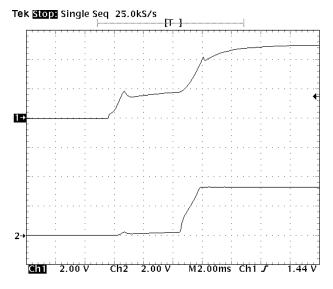


Figure 8 - Start-up at lout=1A. Ch1: 5V input voltage, 2V/div Ch2: 3.3V output voltage, 2V/div

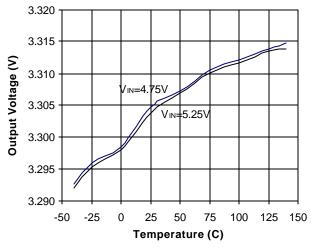


Figure 10 - Thermal shutdown removal response. $I_{LOAD} = 10 mA$ Ch1 Peak: 3.44V

Ch1: Output voltage, 1V/div Ch4: Load Current, 1A/div



Unless specified, the test data applies: T_A=25°C, C_{IN}=4.7μF, C_{OUT}=4.7μF ceramic and V_{IN}=5V.



3.302 3.300 3.298 Output Voltage (V) 3.296 VIN=4.75V 3.294 VIN=5.25V 3.292 3.290 3.288 3.286 3.284 3.282 -25 50 75 100 125 Temperature (C)

Figure 11 - Output Voltage vs. Temperature (Io=5mA).

Figure 12 - Output Voltage vs. Temperature (Io=1A).

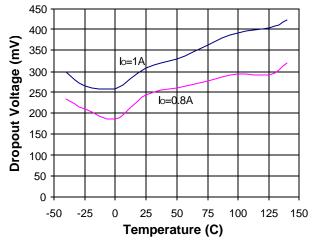


Figure 13 - Dropout Voltage vs. Temperature and Load Current.

Unless specified, the test data applies: T_A=25°C, C_{IN}=4.7μF, C_{OUT}=4.7μF ceramic and V_{IN}=5V.

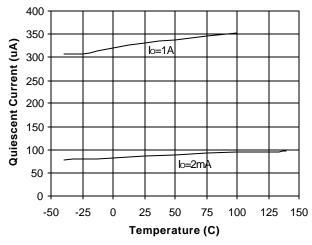


Figure 14 - Quiescent Current vs. Load Current and Temperature.

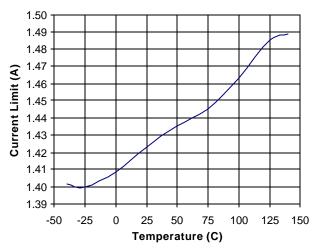


Figure 15 - Typical Current Limit vs. Temperature (V_{IN}=5.5V)

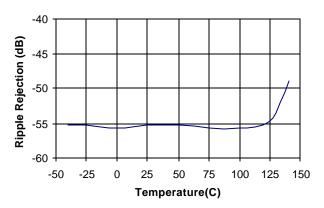


Figure 16 - 120Hz Ripple Rejection vs. Temperature.



APPLICATION INFORMATION

Introduction

The IRU1502-33 regulator is a 3-terminal device offered in a fixed output of 3.3V and it is designed specifically to provide an extremely low dropout voltage.

The IRU1502-33 is designed to meet the fast current transient needs as well as providing an accurate initial voltage, reducing the overall system cost with the need for fewer number of output capacitors.

Thermal Protection

When the junction temperature exceeds 135°C, the internal thermal protection shuts the IRU1502-33 down.

Current Limit Protection

The IRU1502-33 provides Over Current Protection when the output current exceeds typically 1.4A. The output decreases to limit the power dissipation.

Stability

The IRU1502-33 requires the use of an output capacitor as part of the frequency compensation in order to make the regulator stable. A minimum input capacitance of $4.7\mu F$ and a minimum output capacitance $4.7\mu F$ Ceramic capacitor is needed for regulator stage as well as the specified minimum loads to guarantee stability.

Transient Response and PSRR

The input and output capacitors are critical in order to ensure good transient response and PSRR. The most important aspects of this are capacitor selection, placement and trace routing. Place each capacitor as close as physically possible to its corresponding regulator pin. Use wide traces for a low inductance path. Couple directly to the ground and power planes as possible. The use of low ESR capacitors is crucial to achieving good results. Larger capacitance and lower ESR will improve both PSRR and transient response.

Thermal Design

The IRU1502-33 incorporates an internal thermal shutdown that protects the device when the junction temperature exceeds the allowable maximum junction temperature. Although this device can operate with junction temperatures in the range of 150°C, it is recommended that the selected heat sink be chosen such that during maximum continuous load operation the junction temperature is kept below 125°C. The following shows the typical thermal design.

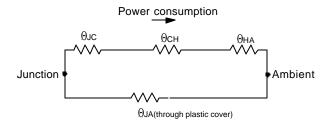


Figure 17 - Thermal resistor diagram for IRU1502-33.

Where:

 Θ_{JC} is the thermal resistance from junction to case. Θ_{CH} is the thermal resistance from case to heat sink if applicable.

 Θ_{HA} is the thermal resistance from heat sink to ambient.

QJA(through plastic cover) is the thermal resistance from junction to the ambient through plastic cover. Typically it is very large and can be neglected. Therefore, overall junction-to-ambient thermal resistance can be represented as:

 Θ JA $\cong \Theta$ JC+ Θ CH+ Θ HA

Where Θ_{JA} is the junction to ambient thermal resistance.



The thermal pad of MLPM is connected to a 1 inch square copper through vias for a four layer PCB board design. From the datasheet, this thermal junction-to-ambient resistance is given as:

 $\Theta_{JA}=42^{\circ}C/W$

Where:

 Θ JC $\cong 2.3^{\circ}C$

 Θ сн \cong 1°С

 $\Theta_{HA} \cong 38.7^{\circ}C$

For IRU1502-33, the thermal design needs to be consider so that the resultant junction temperature is lower than the maximum operating temperature, which is 125°C. Therefore:

$$T_J = \Theta_{JA} \times P_D + T_A \le 125^{\circ}C$$

Assuming, the following conditions:

 $V_{OUT} = 3.3V$

 $V_{IN} = 5V$

lout = 1A (DC Avg)

Calculate the maximum power dissipation using the following equation:

$$P_D = I_{OUT} \times (V_{IN} - V_{OUT})$$

$$P_D = 1 \times (5 - 3.3) = 1.7W$$

For MLPM package, we have:

 $\theta_{JA} = 42^{\circ}C/W$

 $T_A = 45^{\circ}C$

 $\Delta T = P_D \times \theta_{JA} = 1.7 \times 42 = 71.4$

 $T_J = T_A + \Delta T = 116.4^{\circ}C$

Layout Consideration

The IRU1502-33, like many other high-speed regulators, requires that the output capacitors be close to the device for stability. For power consideration, a ground plane pad of approximately one-inch square on the component side must be dedicated to the device where all ground pins are connected to dissipate the power. If a multilayer board is used, it is recommended that the inner layers of the board are also dedicated to the size of the pad for better thermal characteristics.

> International IR Rectifier

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(H) MLPM Package 6-Pin D/2 E/2 _LL3 EXPOSED PAD E2 (OPTIONAL) PIN 1 MARK AREA L2 L1 (SEE NOTE 1) D2 NOTE 2 TOP VIEW PIN NO. 1 **BOTTOM VIEW** PIN NO. 1 S Note 1: Details of pin #1 are optional, but must be located within the zone indicated. The identifier may be molded, marked or A2 metalized features.

Note 2: If L1 Max is not called out, the metalized feature will extend to the exposed pad. Thus L4 does not apply.

SYMBOL	6-PIN 3x3			
DESIG	MIN	NOM	MAX	
Α	0.80	0.90	1.00	
A1	0.00	0.025	0.05	
A2	0.65	0.70	0.75	
A3	0.15	0.20	0.25	
b	0.33	0.35	0.43	
D	3.00 BSC			
D2	1.92	2.02	2.12	
Е	3.00 BSC			
E2	1.11 1.21 1.31			
е		0.95		
L	0.20	0.29	0.45	
L1	0.16	0.24	0.40	
L2			0.125	
L3	0.17		0.30	
L4	0.17			
R	0.127 REF			
S	0°	10°	12°	

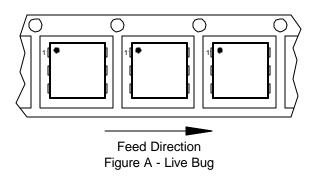
АЗ

SIDE VIEW

NOTE: ALL MEASUREMENTS ARE IN MILLIMETERS.

PACKAGE SHIPMENT METHOD

PKG	PACKAGE	PIN	PARTS	PARTS	T & R
DESIG	DESCRIPTION	COUNT	PER TUBE	PER REEL	Orientation
Н	MLPM 3x3	6		3000	Fig A



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